

ELC1010

Electric Circuits (I) – Circuit Theory



Part (II) – AC Circuits Lecture (2)

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Outline



- V/I Phasor Relationship for R,L and C components.
- Method of solutions in phasor domain
- KCL, KVL in Phasor domain
- Parallel/series combinations
- Voltage/current divider
- Star/delta conversion
- Examples



Circuit elements and phasors



- We need to find out how R, L, and C affect the magnitude and phase relationships between **V** and **I**.
- Once we do this, maybe we can rewrite ohm's law in the complex domain and do everything we did for DC exactly the same way
 - This is valid ONLY for steady-state sinusoid stimulus
- But when we rewrite ohm's law, maybe we will find a new definition for the constant of proportionality
 - This defines imepdances and reactances



Phasor Relationship for Resistor



• Resistor current:

•
$$i = I_m \cos(\omega t + \Phi) = Re\{I_m e^{j(\omega t + \Phi)}\}$$

•
$$I = I_m \angle \Phi$$

• Resistor voltage:

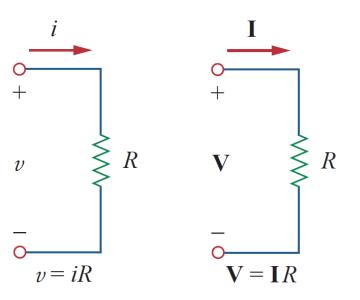
•
$$v = iR = RI_m \cos(\omega t + \Phi)$$

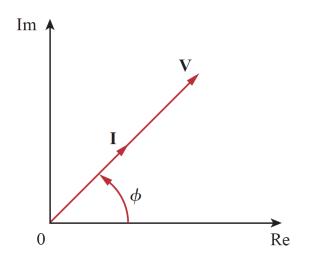
•
$$v = R \times Re\{I_m e^{j(\omega t + \phi)}\}$$

•
$$V = RI_m \angle \Phi$$

• Phasor relation:

•
$$V = RI$$







Phasor Relationship for Inductor



• Inductor current:

•
$$i = I_m \cos(\omega t + \Phi) = Re\{I_m e^{j(\omega t + \Phi)}\}$$

•
$$I = I_m \angle \Phi$$

• Inductor voltage:

•
$$v = L \frac{di}{dt} = -\omega L I_m \sin(\omega t + \Phi)$$

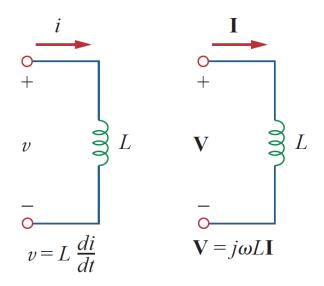
•
$$v = \omega L I_m \cos(\omega t + \Phi + 90^\circ)$$

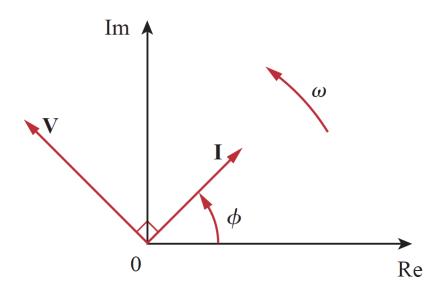
•
$$v = Re\{\omega LI_m e^{j(\omega t + \phi + 90^o)}\}$$

•
$$V = \omega L I_m \angle (\Phi + 90^\circ) = j\omega L I_m \angle \Phi$$

• Phasor relation:

•
$$V = j\omega L \times I$$







Phasor Relationship for Capacitor



Capacitor voltage:

•
$$v = V_m \cos(\omega t + \Phi) = Re\{V_m e^{j(\omega t + \Phi)}\}$$

•
$$V = V_m \angle \Phi$$

• Capacitor current:

•
$$i = C \frac{dv}{dt} = -\omega C V_m \sin(\omega t + \Phi)$$

•
$$i = \omega CV_m \cos(\omega t + \Phi + 90^\circ)$$

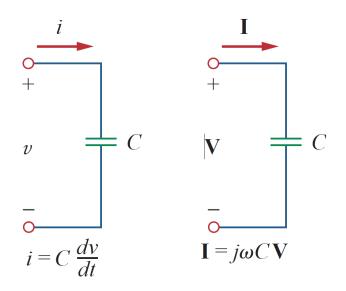
•
$$i = Re\{\omega CV_m e^{j(\omega t + \phi + 90^o)}\}$$

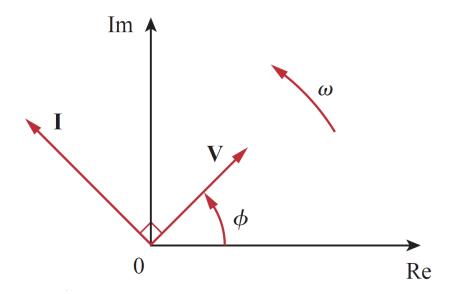
•
$$I = \omega C V_m \angle (\Phi + 90^\circ) = j\omega C V_m \angle \Phi$$

Phasor relation:

•
$$I = j\omega C \times V \rightarrow V = \frac{1}{j\omega C}I$$

• *I* leads *V* by 90°







Summary of voltage-current relationships



Element	Time domain	Frequency domain
R	v = Ri	$\mathbf{V} = R\mathbf{I}$
L	$v = L \frac{di}{dt}$	$\mathbf{V} = j\omega L\mathbf{I}$
C	$i = C \frac{dv}{dt}$	$\mathbf{V} = \frac{\mathbf{I}}{j\omega C}$



Impedance and Admittance



- The impedance Z of a circuit is the ratio of the phasor voltage V to the phasor current I, measured in ohms (Ω)
- Ohm's law is now V = IZ, always and for everyone
- Just calculate the right Z



Impedance and Admittance



• The admittance Y of a circuit is the ratio of the phasor current I to the phasor voltage V, measured in siemens (S or Ω^{-1})

$$I = V/R$$

$$I = \frac{V}{j\omega L}$$

$$I = j\omega CV$$

$$Y = \frac{1}{R}$$

$$Y = \frac{1}{j\omega L}$$

$$Y = j\omega C$$



Impedance and Admittance



- For a circuit with multiple elements, the equivalent Z can be calculated using parallel and series combinations. It ends up being a complex number.
- Impedance Z = R + jX
 - **R** is the equivalent, while X is the reactance
 - Z is inductive if X is +ve
 - **Z** is capacitive if **X** is –ve
 - Z, R, and X have the units of Ω
- Admittance Y = G + jB
 - G is the conductance, while B is the susceptance
 - Y is inductive if B is -ve
 - Y is capacitive if B is +ve
 - Y, G, and B have the units of Ω^{-1}



Notes about impedance



- Resistive impedance is real
 - Resistance is the only thing that dissipates power
- Capacitive and inductive impedance is imaginary
 - They store power
- What part of a complex Z dissipates power and what part stores?
- A purely resistive impedance will not change phase



Impedance and frequency



- Resistive impedance $Z_R = R$ is not a function of frequency
- Capacitance impedance is $Z_C = 1/j\omega C$
- So is inductive impedance $Z_L = j\omega L$
- At DC $\omega = 0$
 - $Z_C = \frac{1}{i\omega C} = \infty \rightarrow \text{Capacitors are O.C.}$
 - $Z_L = j\omega L = 0 \rightarrow \text{Inductors are S.C.}$
 - $Z_R = R \rightarrow$ Only resistors remain
- At very high frequencies $\omega \to \infty$
 - $Z_C = \frac{1}{i\omega C} = 0 \rightarrow \text{Capacitors are S.C.}$
 - $Z_L = j\omega L = \infty$ Inductors are O.C.
 - $Z_R = R \rightarrow$ Only resistors remain
- This is the basis of filters



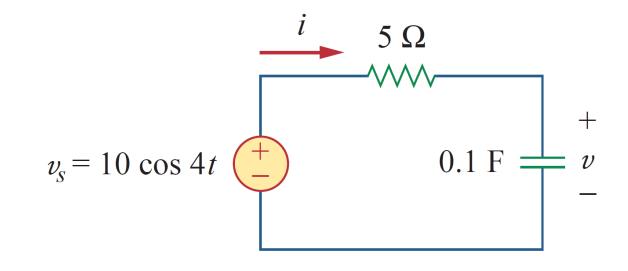


- For the circuit shown find v(t) and i(t)
- $V_S = 10 \angle 0^\circ$, $\omega = 4rad/sec$

•
$$Z = R + \frac{1}{j\omega C}$$

•
$$Z = 5 - j2.5 \Omega$$

•
$$I = \frac{V_S}{Z} = \frac{10}{5 - j2.5}$$



•
$$V = \frac{10}{5-j2.5} \times (-j2.5) = 10 \times \frac{-j2.5}{5-j2.5}$$
 (potential divider)



Note:



- For any complex number z = x + jy,
- The conjugate is $z^* = x jy$

•
$$z \times z^* = (x + jy)(x - jy) = x^2 - jxy + jxy - j^2y^2$$

• *Hence*,
$$z \times z^* = x^2 + y^2 = |z|^2$$





•
$$I = \frac{10}{5-j2.5} \frac{5+j2.5}{5+j2.5}$$

•
$$I = \frac{50+j25}{31.25} = 1.6 + j0.8$$

•
$$I = 1.79 \angle 26.57^{\circ} A$$

•
$$i(t) = 1.79 \cos(4t + 26.57^{\circ}) A$$

•
$$V = IZ_C = \frac{I}{j\omega C} = \frac{1.79 \angle 26.57^{\circ}}{j0.4} = \frac{1.79 \angle 26.57^{\circ}}{0.4 \angle 90}$$

•
$$V = 4.47 \angle - 63.43^{\circ}$$

•
$$v(t) = 4.47 \cos(4t - 63.43^{\circ})$$



Summary of AC Circuits



- Voltage and current and written as magnitude and phase
- $v(t) = V_m \cos(\omega t + \phi) = Re(V_m e^{j\omega t + j\phi})$
 - $V = V_m \angle \phi$
- $i(t) = I_m \cos(\omega t + \theta) = Re(I_m e^{j\omega t + j\theta})$
 - $I = I_m \angle \theta$
- Impedance is defined as $Z = \frac{V}{I}$
 - For resistors Z = R
 - For inductors $Z = j\omega L$
 - For capacitors $Z = \frac{1}{j\omega C}$



Kirchhoff's Laws in the Phasor Domain



KVL and KCL are valid in phasor domain

- KVL
 - Let $v_1, v_2, ..., v_n$ be the voltages around a closed loop.
 - Then:

$$v_1 + v_2 + \dots + v_n = 0$$

• In the sinusoidal steady state, each voltage may be written in cosine form

$$V_{m1}\cos(\omega t + \Phi_1) + V_{m2}\cos(\omega t + \Phi_2) + \cdots + V_{mn}\cos(\omega t + \Phi_n) = 0$$



Kirchhoff's Laws in the Phasor Domain



Then

$$Re\{V_{m1}e^{j\Phi_1}e^{j\omega t}+V_{m2}e^{j\Phi_2}e^{j\omega t}+\cdots+V_{mn}e^{j\Phi_n}e^{j\omega t}\}=0$$

$$Re\{V_{m1}e^{j\Phi_1}\} + Re\{V_{m2}e^{j\Phi_2}\} + \cdots + Re\{V_{mn}e^{j\Phi_n}\} = 0$$

• Let

$$V_k = V_{mk}e^{j\Phi_k}$$

Then

$$\boldsymbol{V}_1 + \boldsymbol{V}_2 + \dots + \boldsymbol{V}_n = 0$$

So, Kirchhoff's voltage law holds for phasors



Kirchhoff's Laws in the Phasor Domain



• Similarly, Kirchhoff's current law holds for phasors at all nodes

$$I_1 + I_2 + \dots + I_n = 0$$



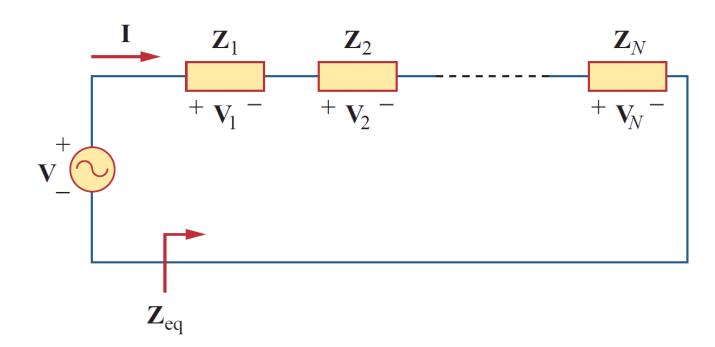
Impedance Combination



 For N series connected impedances, the same current I flows through the impedances

•
$$V = V_1 + V_2 + \dots + V_N = I(Z_1 + Z_2 + \dots + Z_N)$$

•
$$Z_{eq} = \frac{V}{I} = Z_1 + Z_2 + \dots + Z_N$$

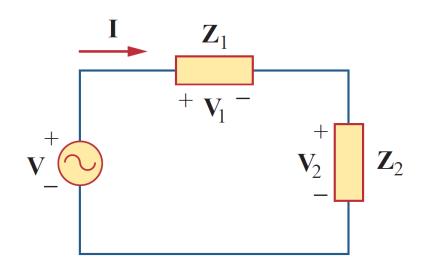




Voltage Divider



- For two impedances connected in series
 - $\bullet \ I = \frac{V}{Z_1 + Z_2}$
 - $V_1 = IZ_1, V_2 = IZ_2$
 - $V_1 = \frac{Z_1}{Z_1 + Z_2} V$, $V_2 = \frac{Z_2}{Z_1 + Z_2} V$





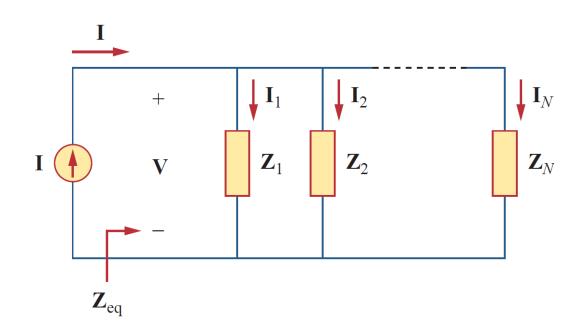
Admittance Combination



- For N Parallel connected impedance, voltage is the same across all impedances
- Applying KCL

•
$$I = I_1 + I_2 + \dots + I_N = V\left(\frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_N}\right)$$

- $\frac{1}{Z_{eq}} = \frac{I}{V} = \frac{1}{Z_1} + \frac{1}{Z_2} + \dots + \frac{1}{Z_N}$
- $Y_{eq} = Y_1 + Y_2 + \cdots + Y_N$





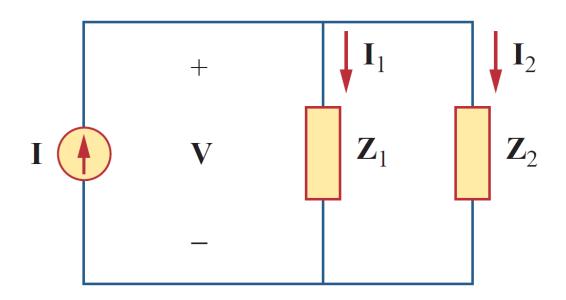
Current Divider



•
$$Z_{eq} = \frac{1}{Y_{eq}} = \frac{1}{Y_{1} + Y_{2}} = \frac{1}{1/Z_{1} + 1/Z_{2}} = \frac{Z_{1}Z_{2}}{Z_{1} + Z_{2}}$$

•
$$V = IZ_{eq} = I\frac{Z_1Z_2}{Z_1+Z_2} = I_1Z_1 = I_2Z_2$$

•
$$I_1 = \frac{Z_2}{Z_1 + Z_2} I$$
, $I_2 = \frac{Z_1}{Z_1 + Z_2} I$



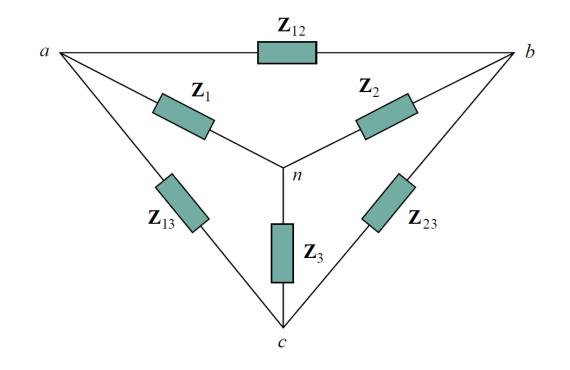


Star-Delta Conversion



•
$$Z_{12} = Z_1 + Z_2 + \frac{Z_1 Z_2}{Z_3}$$

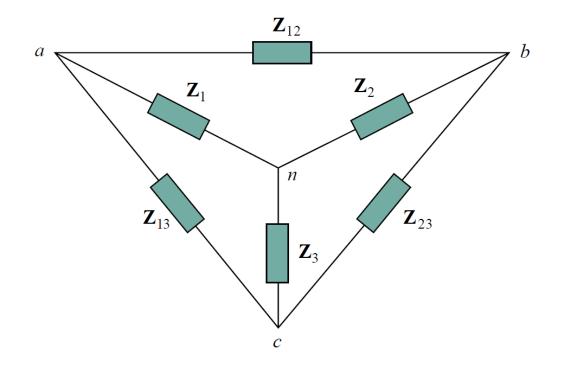
•
$$Z_{23} = Z_2 + Z_3 + \frac{Z_2 Z_3}{Z_1}$$





Delta-Star Conversion

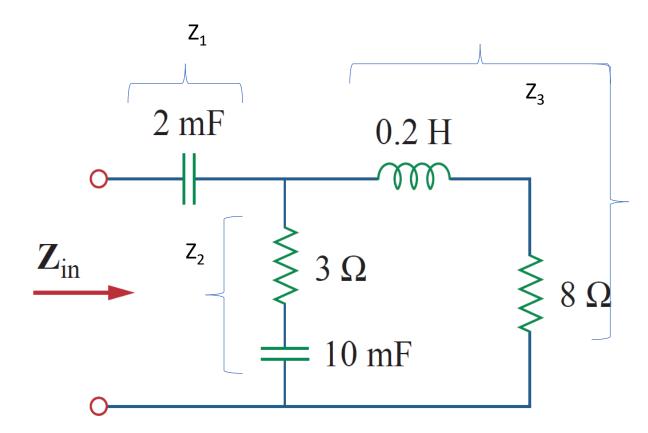








• Find the input impedance of the circuit shown. Assume that the circuit operates at $\omega = 50$ rad/s







• Find the input impedance of the circuit shown. Assume that the circuit

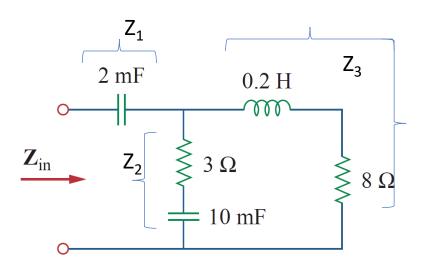
2 mF

operates at $\omega = 50$ rad/s

- $Z_1 = Impedance \ of \ 2mF \ Capacitor$ • $Z_1 = \frac{1}{j\omega C} = -j10$
- $Z_2 = R + \frac{1}{i\omega C} = 3 j2$
- $Z_3 =$ Impedance of the 0.2H inductor in series with the 8Ω resistor
 $Z_3 = 8 + j10$







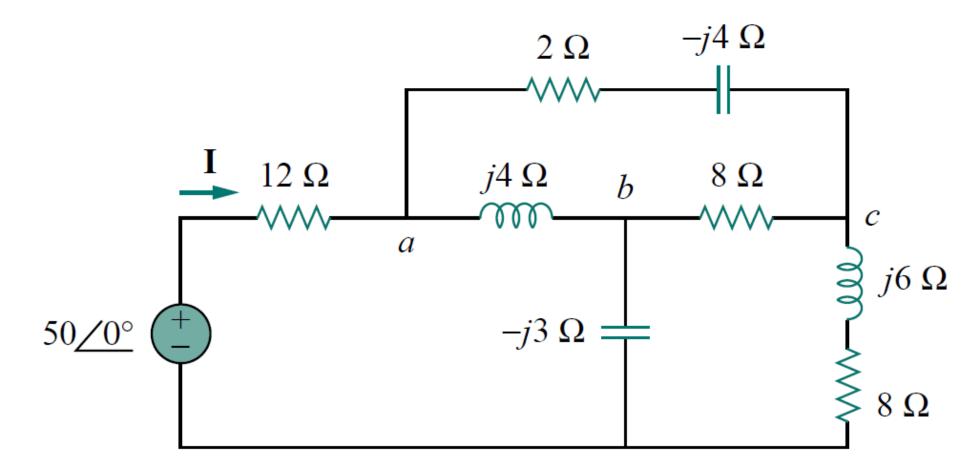
•
$$Z_1 = \frac{1}{j\omega C} = -j10, Z_2 = R + \frac{1}{j\omega C} = 3 - j2, Z_3 = 8 + j10$$

- The input impedance is $Z_{in} = Z_1 + Z_2 || Z_3$
 - $Z_{in} = -j10 + \frac{(3-j2)(8+j10)}{11+j8} = -j10 + \frac{44+j14}{11+j8}$
 - $Z_{in} = -j10 + \frac{(44+j14)(11-j8)}{11^2+8^2} = -j10 + 3.22 j1.07$
 - $Z_{in} = 3.22 j11.07$



Example 2.0

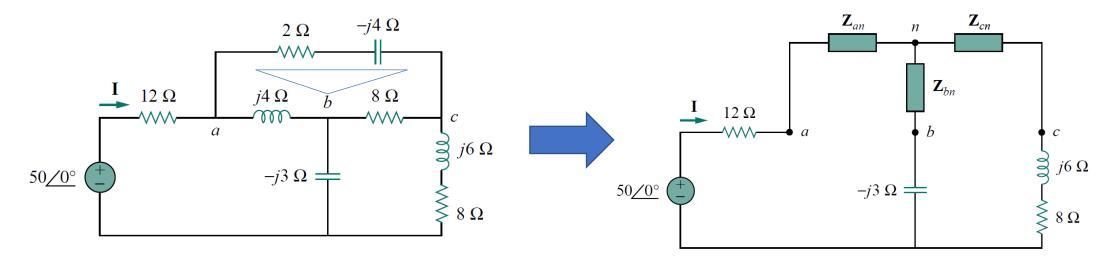






Example 2.0



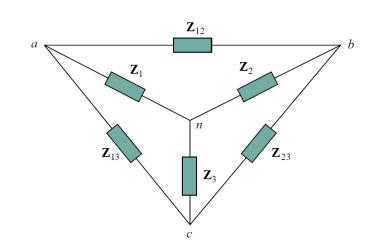


•
$$Z_{cn} = \frac{8(2-j4)}{10} = 1.6 - j3.2$$

•
$$Z_{cn} = \frac{8(2-j4)}{10} = 1.6 - j3.2$$

• $Z_{an} = \frac{j4(2-j4)}{10} = 1.6 + j0.8$
• $Z_{bn} = \frac{j4\times8}{10} = j3.2$

•
$$Z_{bn} = \frac{j4 \times 8}{10} = j3.2$$



$$Z_{1} = \frac{Z_{12}Z_{13}}{Z_{12} + Z_{13} + Z_{23}}$$

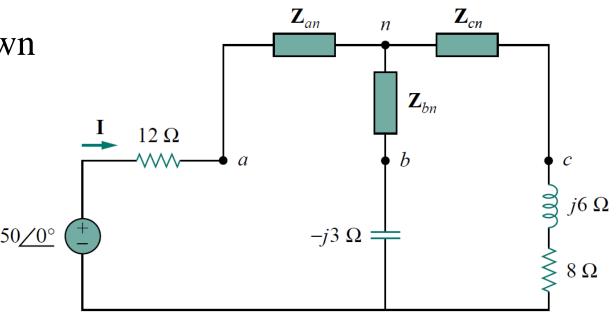
$$Z_{2} = \frac{Z_{12}Z_{23}}{Z_{12} + Z_{13} + Z_{23}}$$

$$Z_{3} = \frac{Z_{13}Z_{23}}{Z_{12} + Z_{13} + Z_{23}}$$



Example 2.0





•
$$Z = 12 + Z_{an} + (Z_{bn} - j3)||(Z_{cn} + 8 + j6)||$$

•
$$Z = 12 + 1.6 + j0.8 + (j0.2)||(9.6 + j2.8)|$$

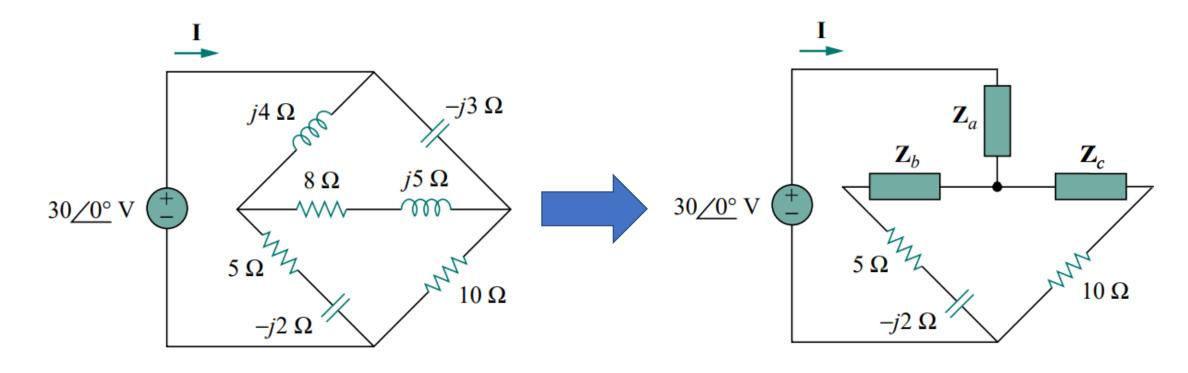
•
$$Z \approx 13.6 + j1 = 13.64 \angle 4.2^{\circ}$$

•
$$I = \frac{V}{Z} = \frac{50 \angle 0^{\circ}}{13.64 \angle 4.2^{\circ}} = 3.67 \angle -4.2^{\circ}$$



Example 3.0





•
$$I = 6.36 \angle 3.8^{\circ}$$