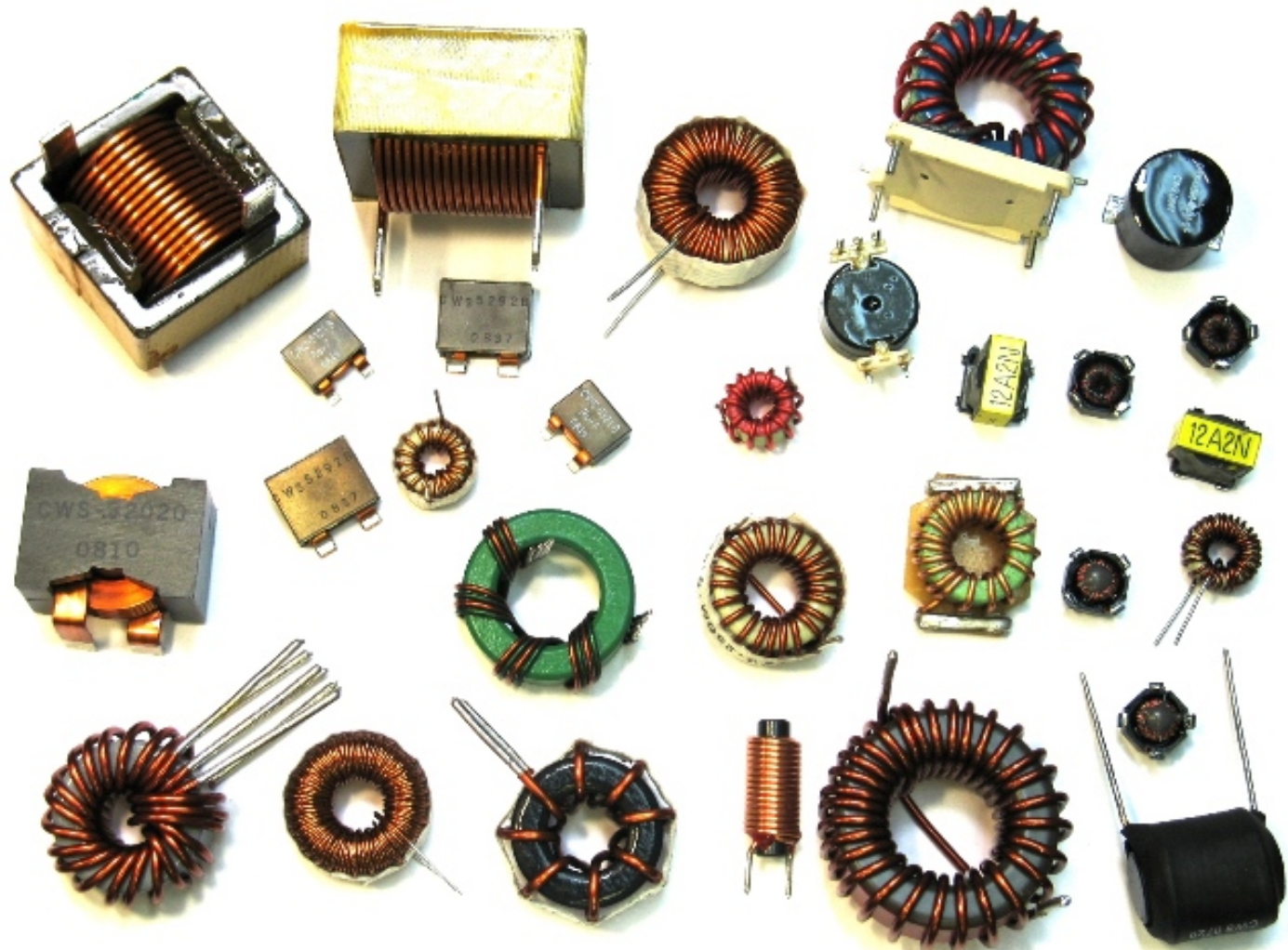


Faraday's Law Applications

1

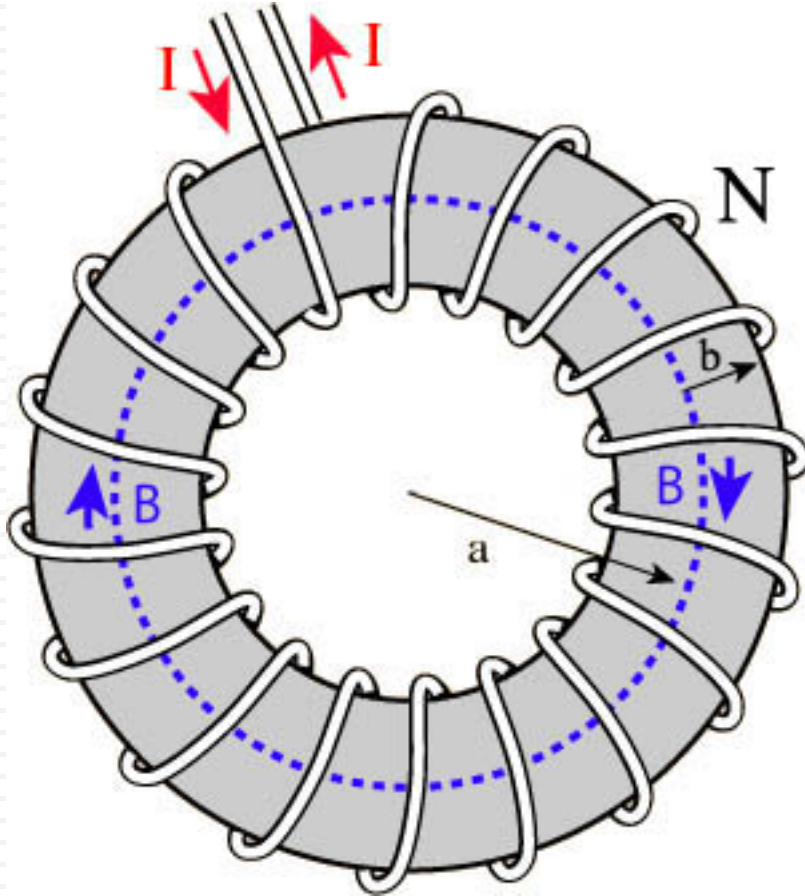
1. Inductors.
2. Transformers.
3. Electrical Generators.
4. Electrical Motors.
5. Transmission Lines.

2

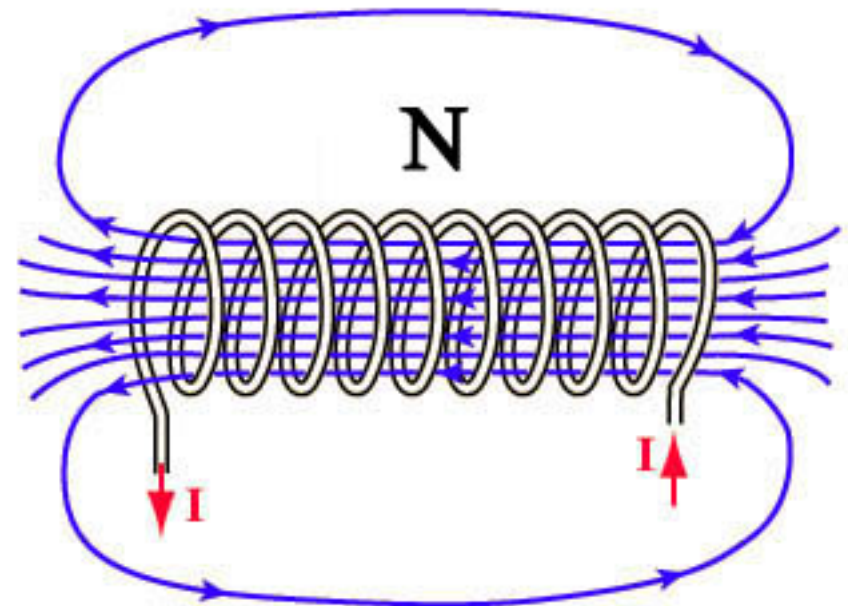


Inductors

3



Toroid

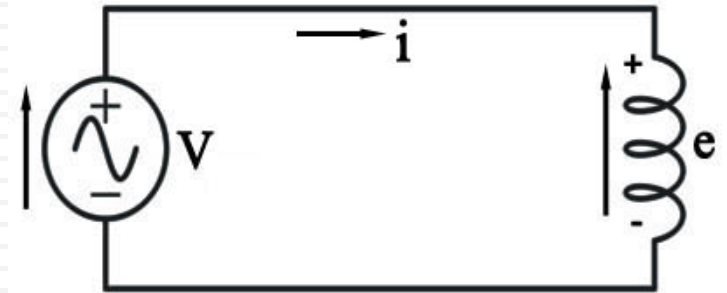


Solenoid

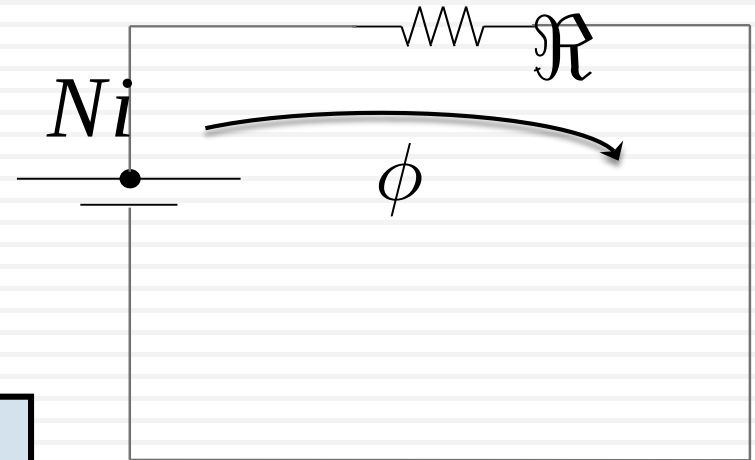
Inductors

4

$$v = N \frac{d\phi}{dt}$$



$$Ni = \phi \mathfrak{R} \quad \Rightarrow \quad \phi = \frac{Ni}{\mathfrak{R}}$$



$$v = \frac{N^2}{\mathfrak{R}} \frac{di}{dt}$$

$$v = L \frac{di}{dt}$$

$$L = \frac{N^2}{\mathfrak{R}}$$

Inductors

5

By Integration

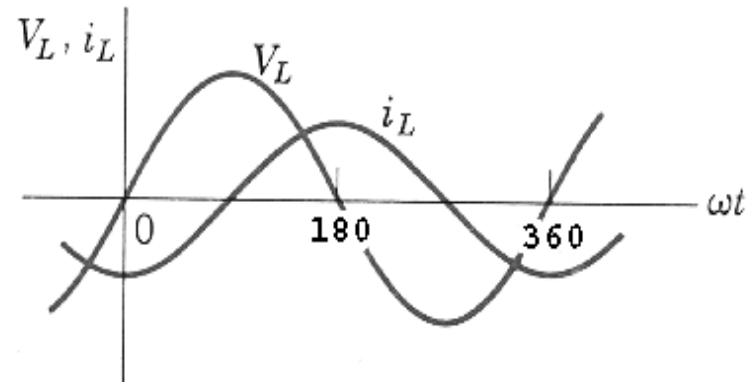
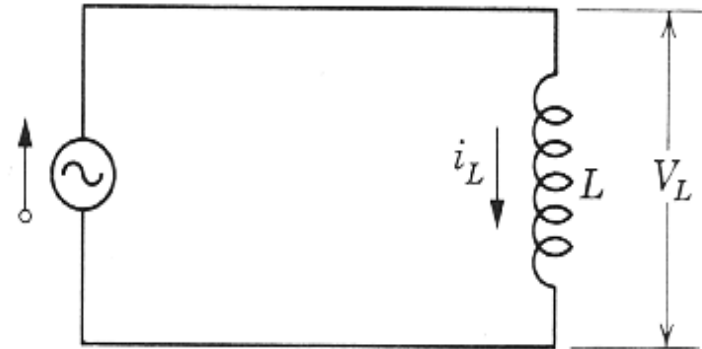
$$i = \frac{V_m}{\omega L} \sin\left(\omega t - \frac{\pi}{2}\right)$$

$$I = \frac{V}{j\omega L}$$



$$V = j(\omega L)I$$

$$V = jX_L I$$

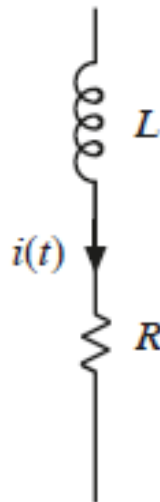


Inductors: Introduction to Inductor Design

6

Objective:

Design an inductor having a given inductance L , which carries worst-case current $I_{\max}$ without saturating the core ($B_{\max}$), and which has a given winding resistance R , or, equivalently, exhibits a worst-case copper loss of $I_m^2 R$



Transformer

7

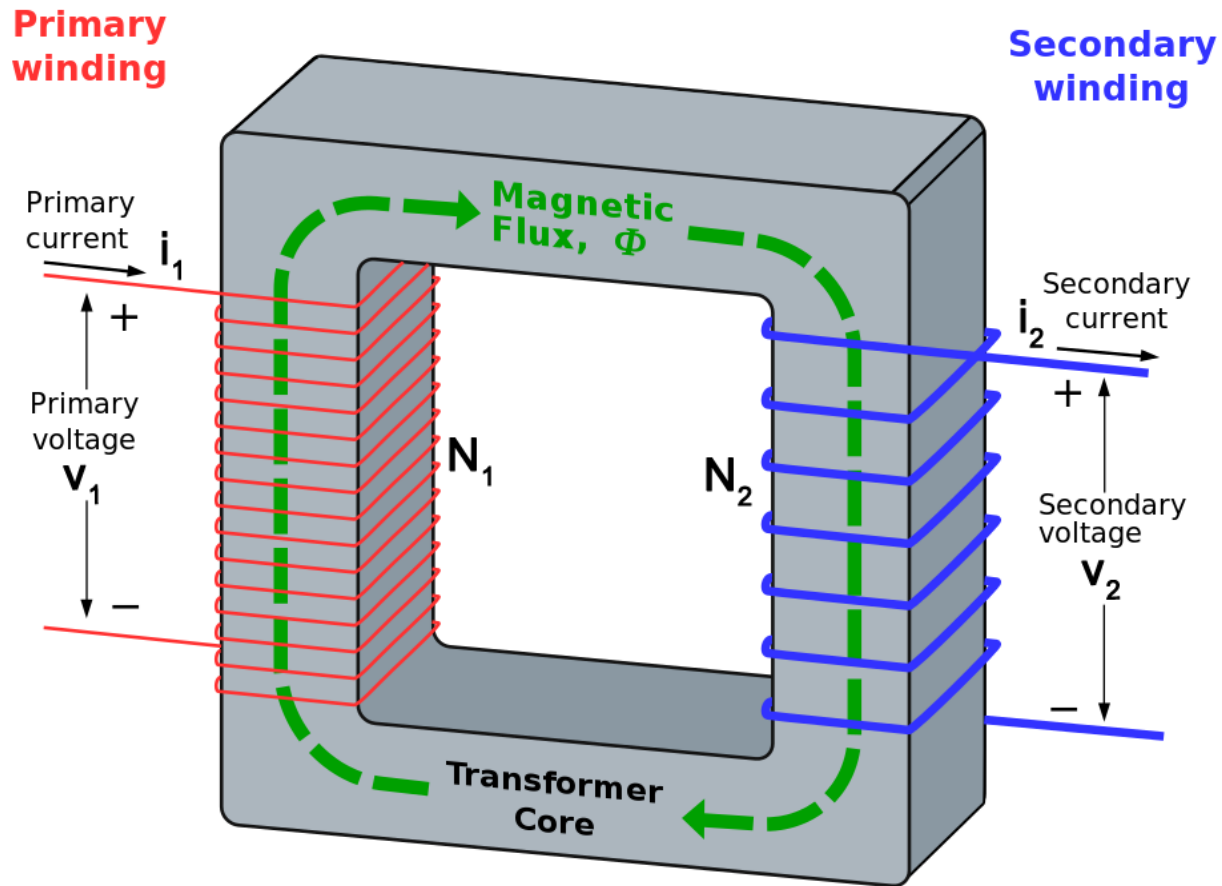
Single-phase transformer



Transformer

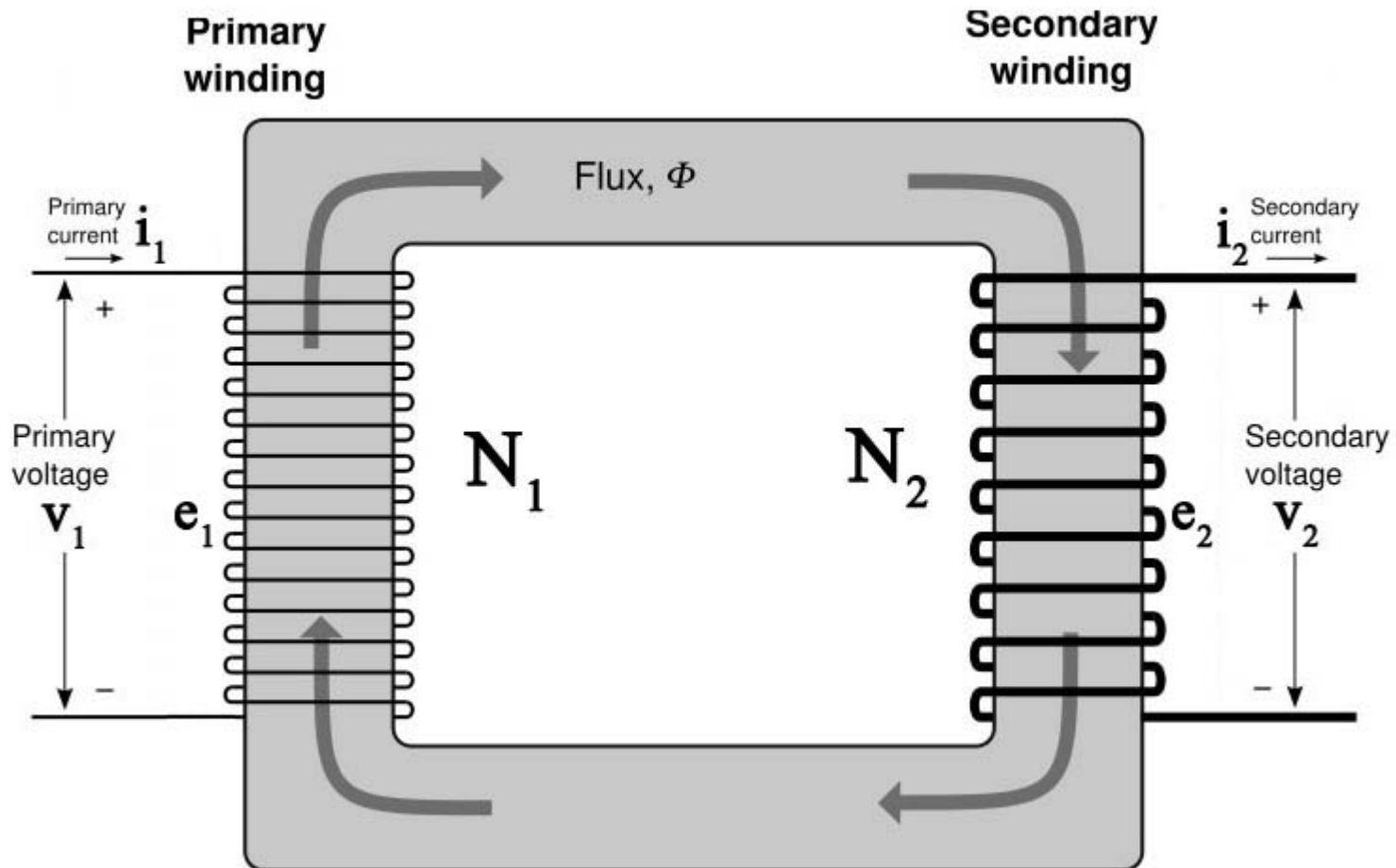
8

Single-phase transformer



Transformer

9



Transformer

10

Ideal Transformer

- $R_{\text{coil}} = \text{zero}$
- $\mu \rightarrow \infty$
- No leakage flux
- No core losses

$$v_1 = N_1 \frac{d\phi_1}{dt}$$

$$v_2 = N_2 \frac{d\phi_2}{dt}$$

Since no leakage

$$\phi_1 = \phi_2$$

$$\frac{v_1}{v_2} = \frac{N_1}{N_2}$$

$$\frac{e_1}{e_2} = \frac{N_1}{N_2}$$

Transformer

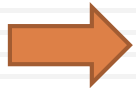
11

Ideal Transformer

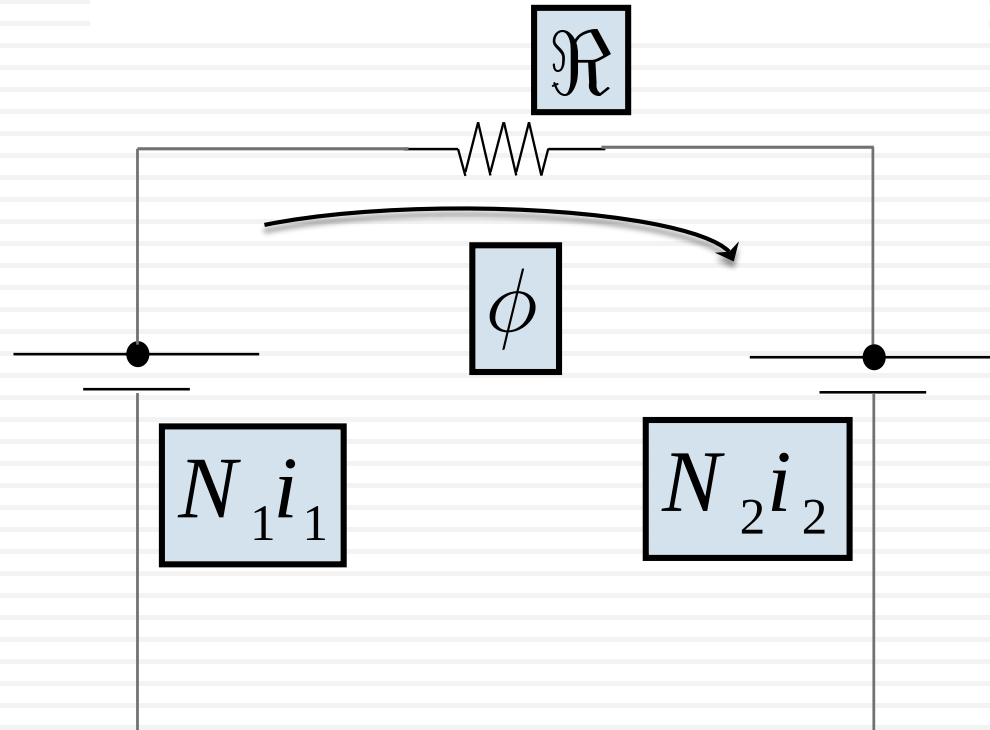
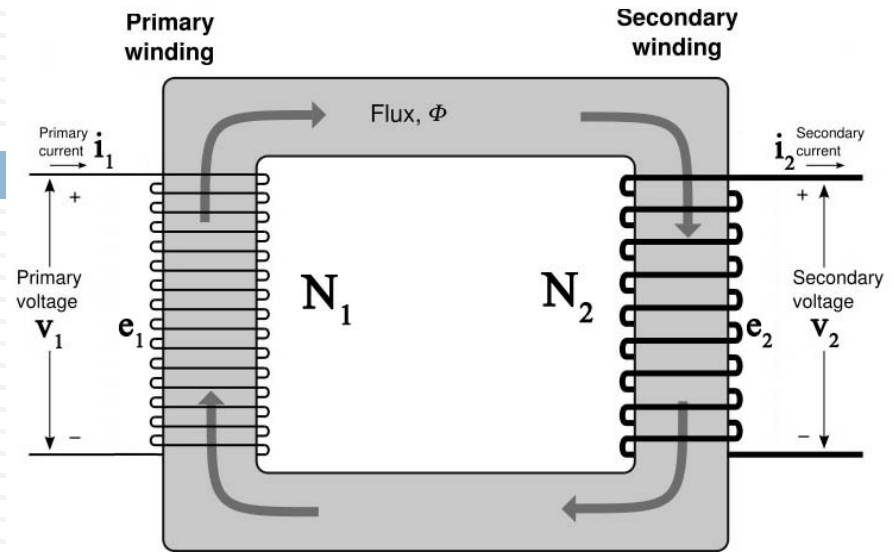
$$N_1 i_1 - N_2 i_2 = \phi \mathcal{R}$$

Since $\mu \rightarrow \infty$

$$\mathcal{R} = \text{zero}$$



$$\frac{N_1}{N_2} = \frac{i_2}{i_1}$$



Transformer

12

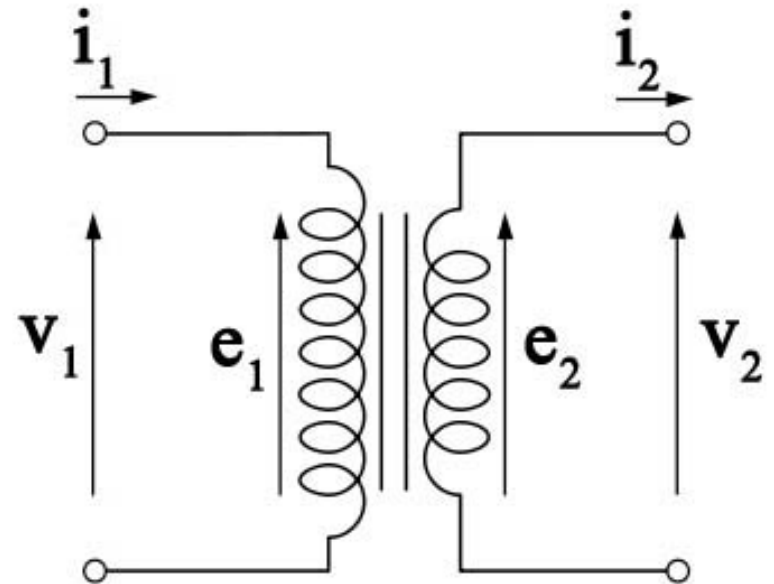
Ideal Transformer

$$\frac{v_1}{v_2} = \frac{N_1}{N_2}$$

$$\frac{e_1}{e_2} = \frac{N_1}{N_2}$$

$$\frac{i_2}{i_1} = \frac{N_1}{N_2}$$

$$v_1 i_1 = v_2 i_2$$



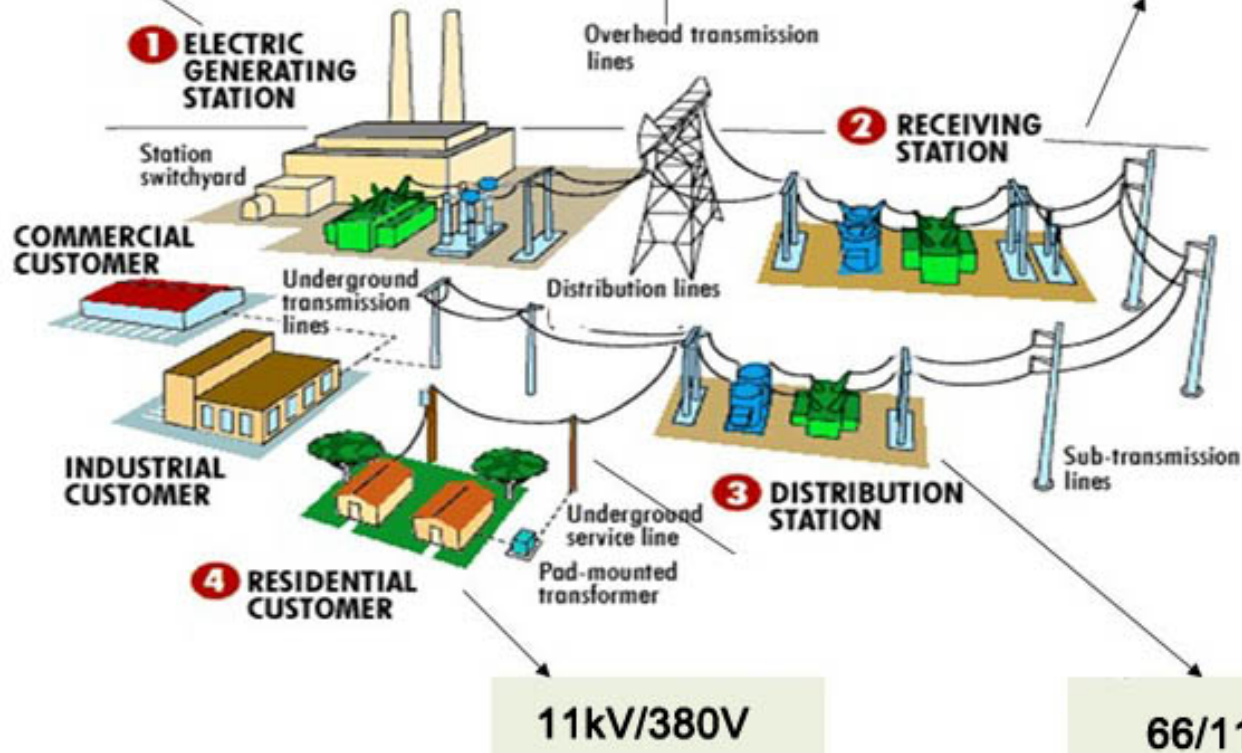
Transformer: Applications

13

3x375 MW
11/13.2/15/18 kV

500 kV
220 kV

220/66 kV



Transformer: Equivalent Circuit

14

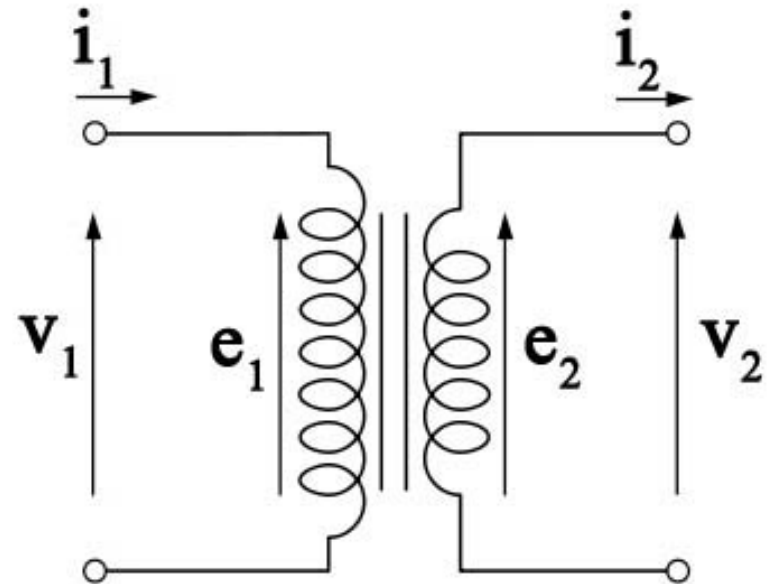
Case A: Ideal Transformer

$$\frac{v_1}{v_2} = \frac{N_1}{N_2}$$

$$\frac{e_1}{e_2} = \frac{N_1}{N_2}$$

$$\frac{i_2}{i_1} = \frac{N_1}{N_2}$$

$$v_1 i_1 = v_2 i_2$$



Transformer: Equivalent Circuit

15

Case B: - $R_{\text{coil}} = \text{zero}$

- $\mu \neq \infty$

- No leakage flux

- No core losses

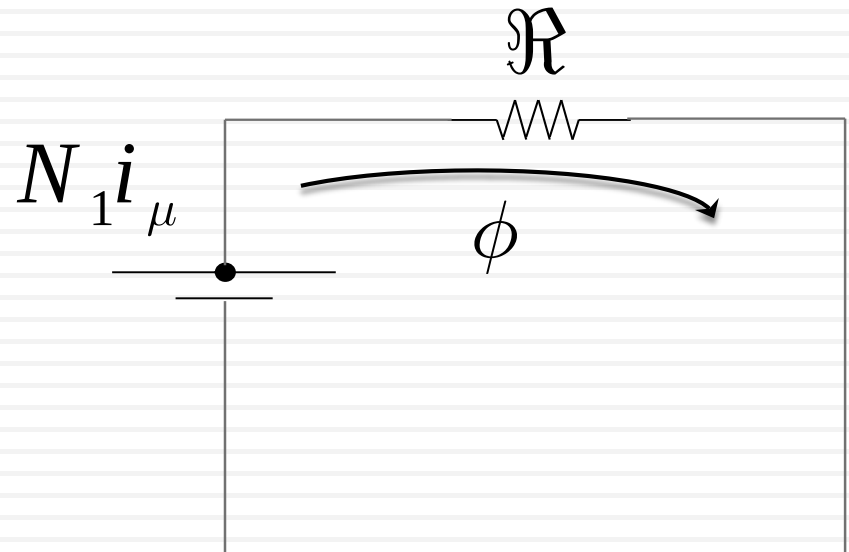
At No-load: ($i_2 = \text{zero}$)

$$N_1 i_\mu = \phi \mathfrak{R}$$

From Faraday's Law

$$\phi = \frac{V_{1m}}{\omega N_1} \sin\left(\omega t - \frac{\pi}{2}\right)$$

$$i_\mu = \frac{V_{1m}}{\omega N_1^2 \mathfrak{R}} \sin\left(\omega t - \frac{\pi}{2}\right)$$

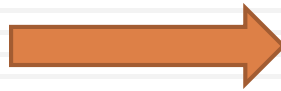


Transformer: Equivalent Circuit

16

$$i_{\mu} = \frac{V_{1m}}{\omega L_{core}} \sin\left(\omega t - \frac{\pi}{2}\right)$$

$$I_{\mu} = \frac{V_1}{j \omega L_{core}}$$

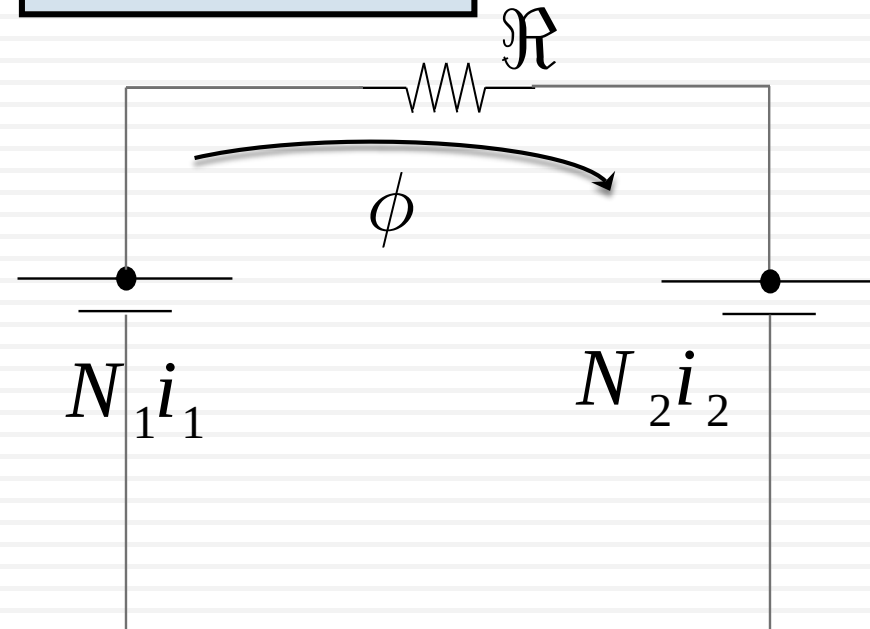


$$X_{\mu} = \omega L_{core}$$

If the transformer is loaded:

$$N_1 i_1 - N_2 i_2 = \phi \mathcal{R}$$

$$N_1 i_1 - N_2 i_2 = N_1 i_{\mu}$$



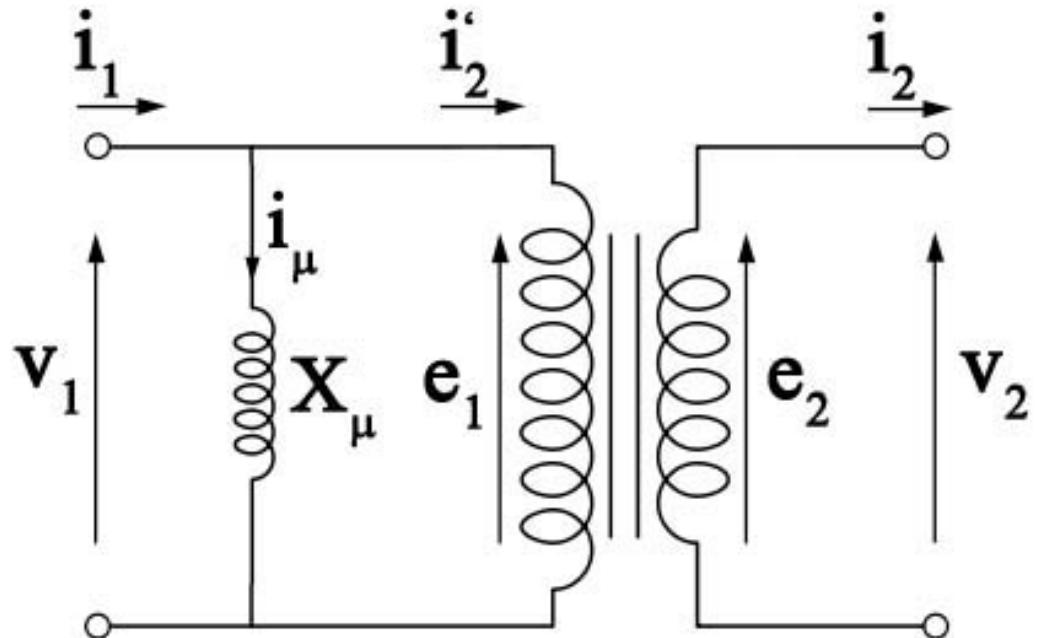
Transformer: Equivalent Circuit

17

$$i_{\mu} = i_1 - \frac{N_2}{N_1} i_2$$

$$i_1 = i_{\mu} + \frac{N_2}{N_1} i_2$$

$$i_2' = \frac{N_2}{N_1} i_2$$



Transformer: Equivalent Circuit

18

Case C: - $R_{\text{coil}} = \text{zero}$

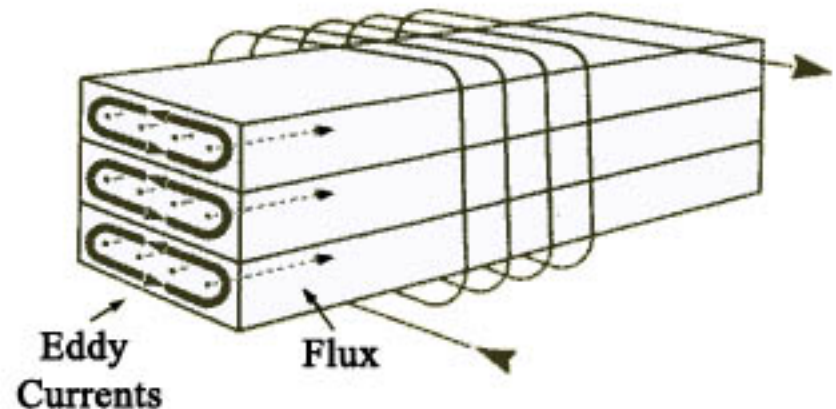
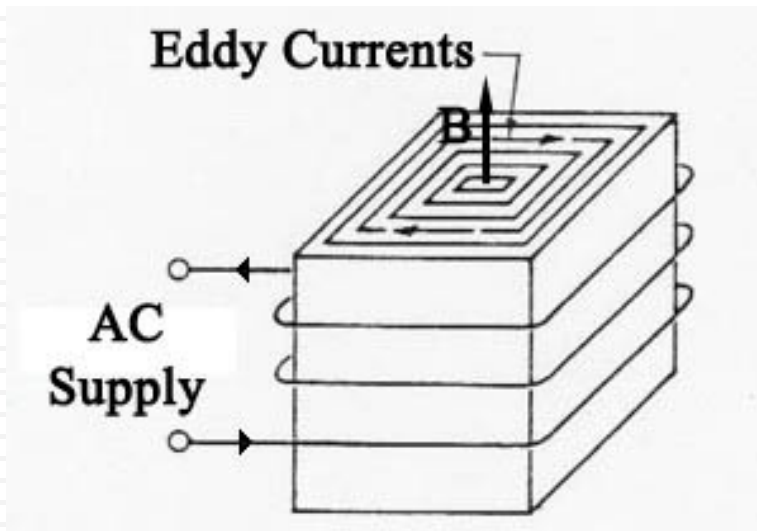
- $\mu \neq \infty$

- No leakage flux

- core losses exist.

Core (iron) losses

1. Eddy Current Losses



$$P_e \propto B_m^2 f^2 (\text{thickness})^2$$

$$P_e = k_e B_m^2 f^2 (\text{thickness})^2$$

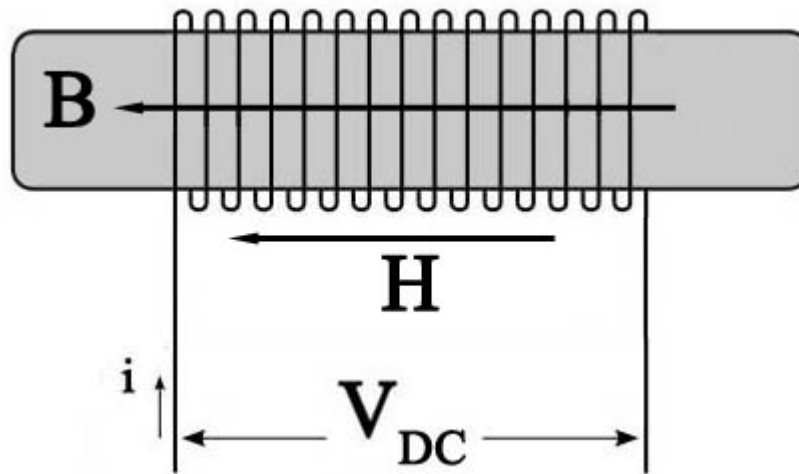
$$W/m^3$$

Transformer: Equivalent Circuit

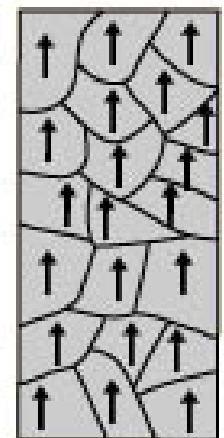
19

Case C:

2. Hysteresis Losses



In bulk material
the domains
usually cancel,
leaving the
material
unmagnetized.



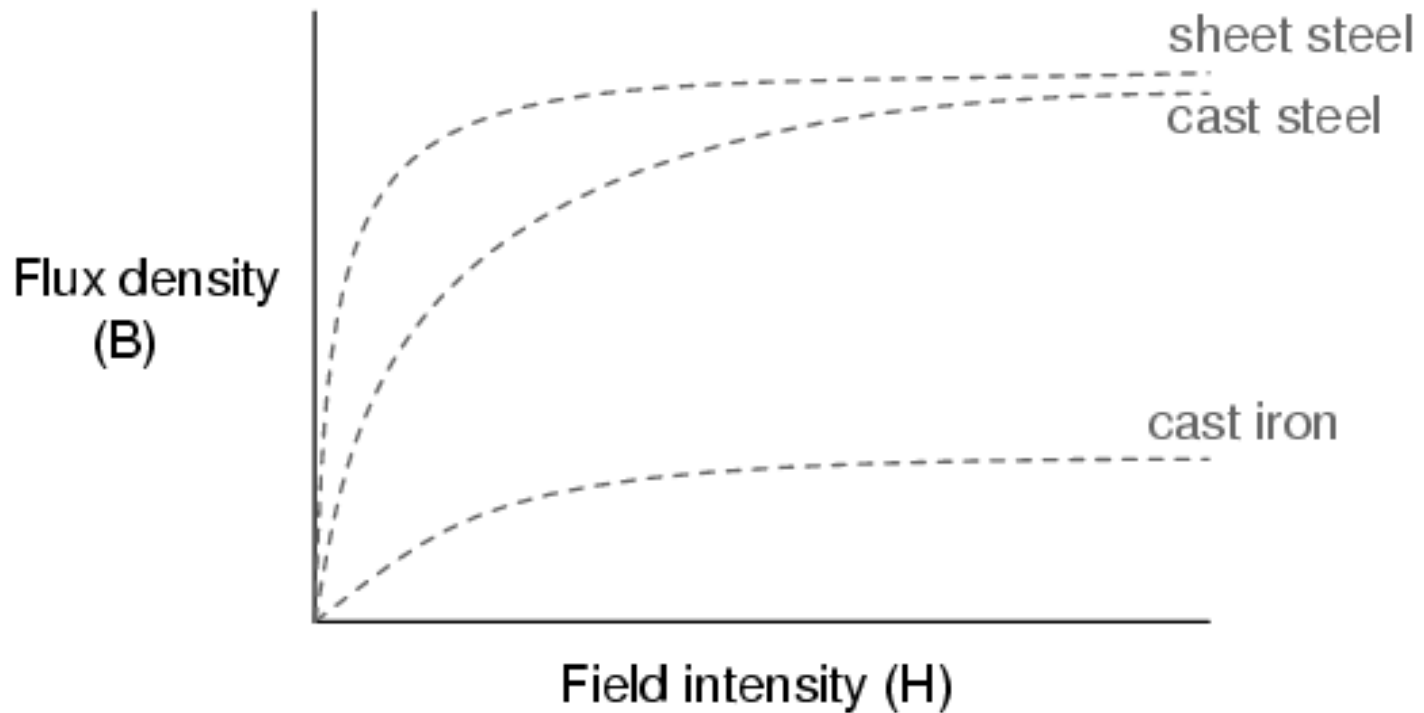
Externally
applied
magnetic field.

Transformer: Equivalent Circuit

20

Case C:

2. Hysteresis Losses



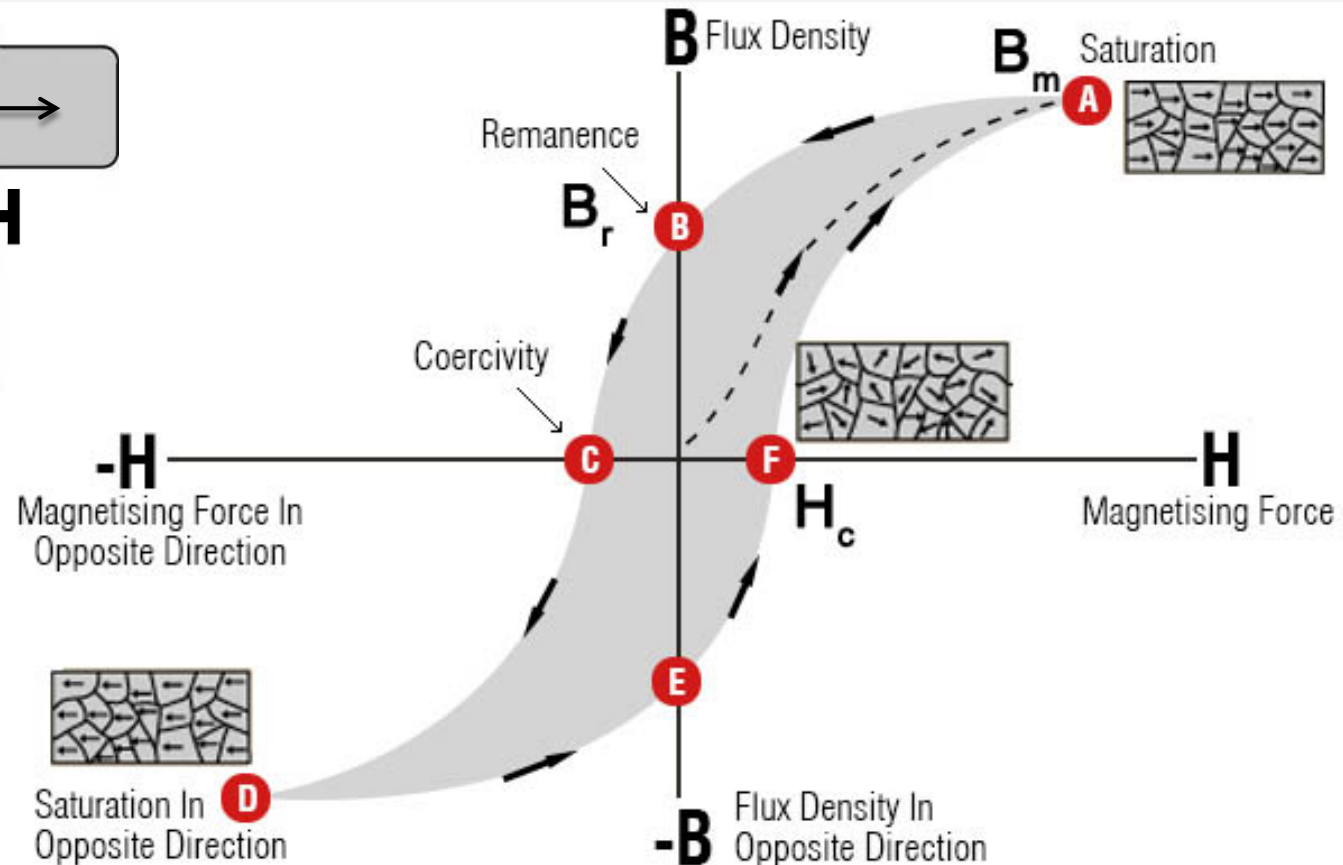
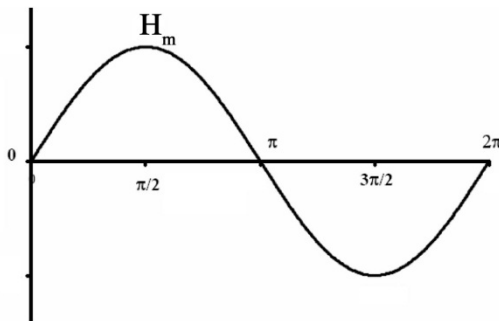
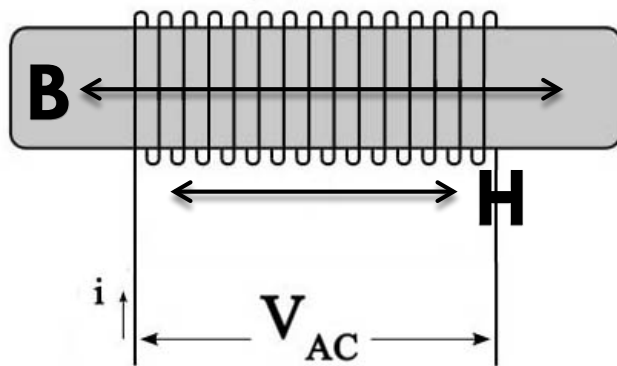
BH Curve (Magnetization Curve)

Transformer: Equivalent Circuit

21

Case C:

2. Hysteresis Losses



Transformer: Equivalent Circuit

22

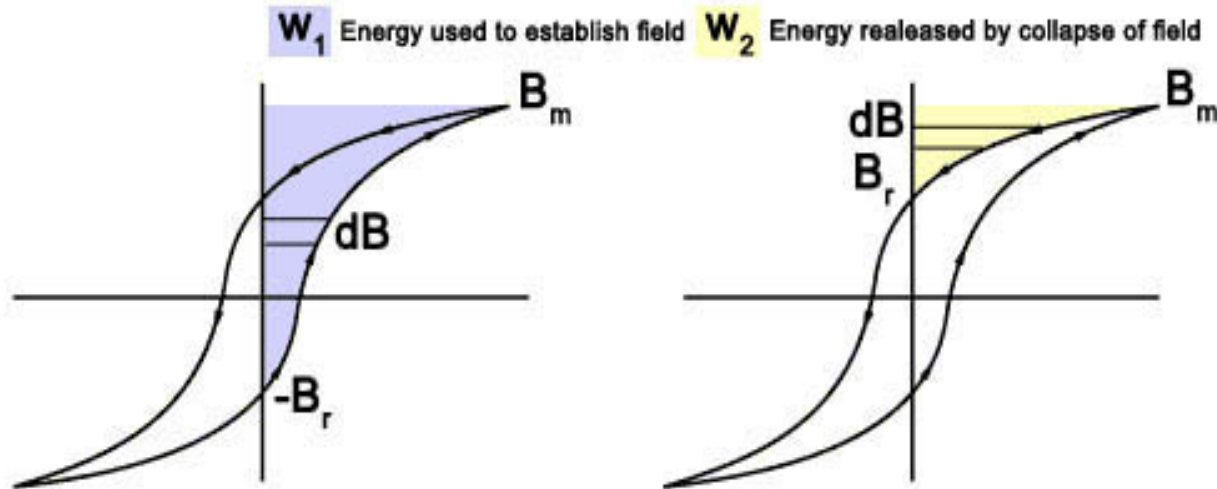
Case C:

2. Hysteresis Losses

$$w_1 = \int_{-B_r}^{B_m} \underline{H} \cdot \underline{dB}$$

$$w_2 = \int_{B_m}^{-B_r} \underline{H} \cdot \underline{dB}$$

$$H \text{ (A.m}^{-1}\text{)} \times B \text{ (T = J.A}^{-1}\text{.m}^{-2}\text{)} = \text{J.m}^{-3}$$



Energy lost /unit volume/cycle = $2(w_1 - w_2)$ = area of the loop

Transformer: Equivalent Circuit

23

Case C:

2. Hysteresis Losses

$$P_h \propto B^n f \text{ (volume)}$$

$$P_h = k_h B^n f$$

$$W/m^3$$

n depends on the material (1.6 – 2.4)

Total core losses:

$$P_c = k_e B^2 f^2 (\text{thickness})^2 + k_h B^n f$$

$$W/m^3$$

Since f is usually constant

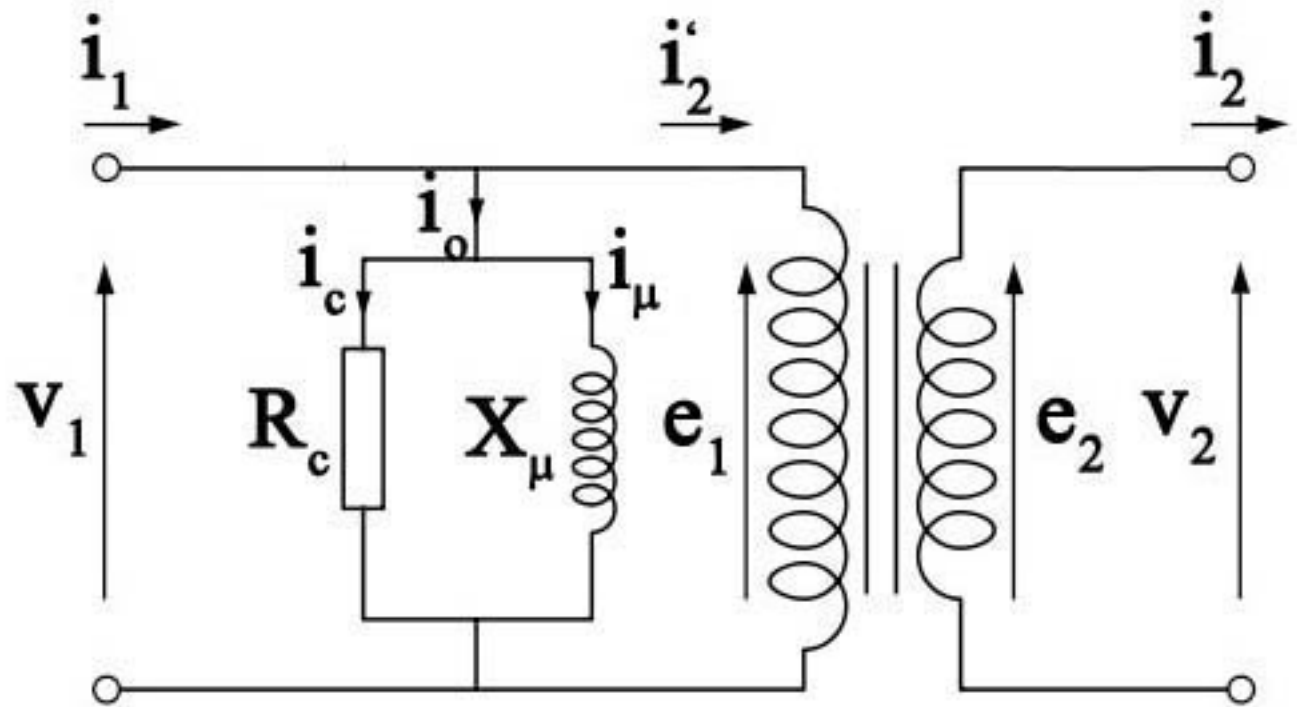
$$\Rightarrow P_c \propto B^2 \Rightarrow P_c \propto \phi^2 \Rightarrow P_c \propto V^2$$

Transformer: Equivalent Circuit

24

Case C:

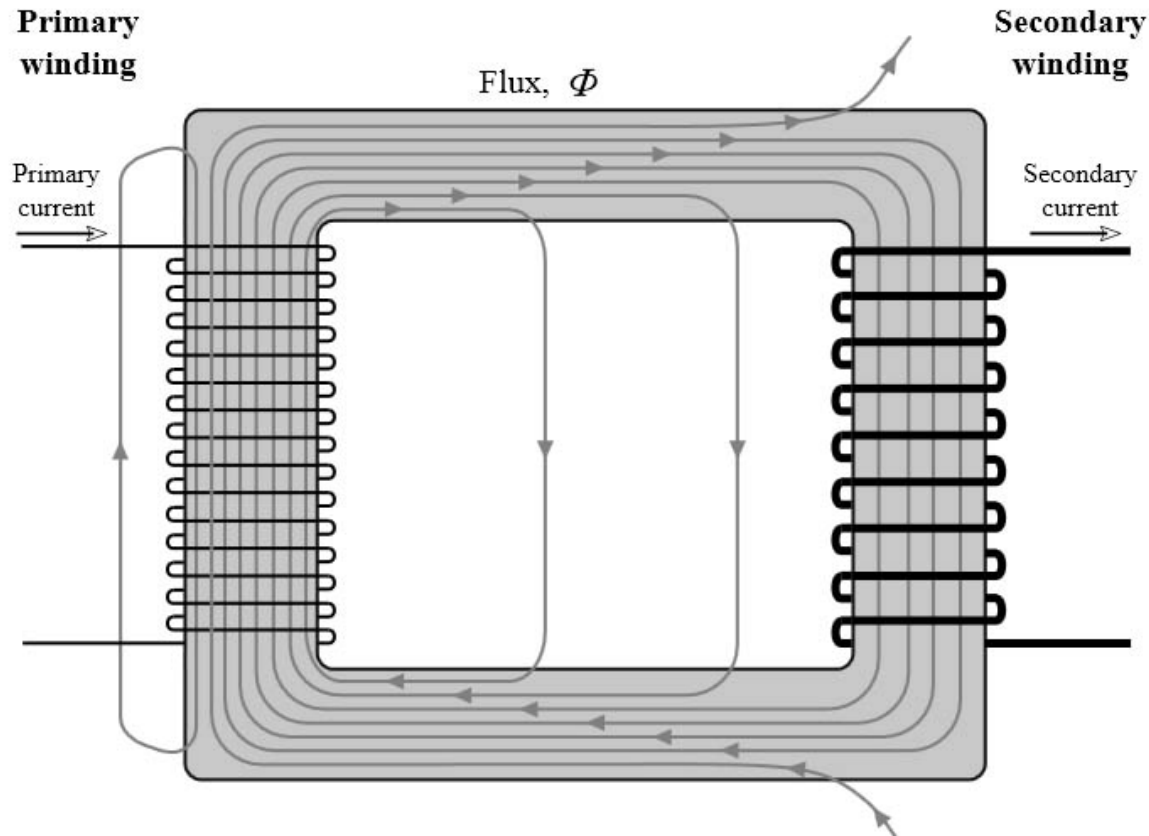
$$P_c = \frac{V_1^2}{R_c}$$



Transformer: Equivalent Circuit

25

- Case D:
- $R_{\text{coil}} = \text{zero}$
 - $\mu \neq \infty$
 - Leakage flux exists.
 - core losses exist.



Transformer: Equivalent Circuit

26

Case D:

$$\phi_1 = \phi + \phi_{l1} \quad (1)$$

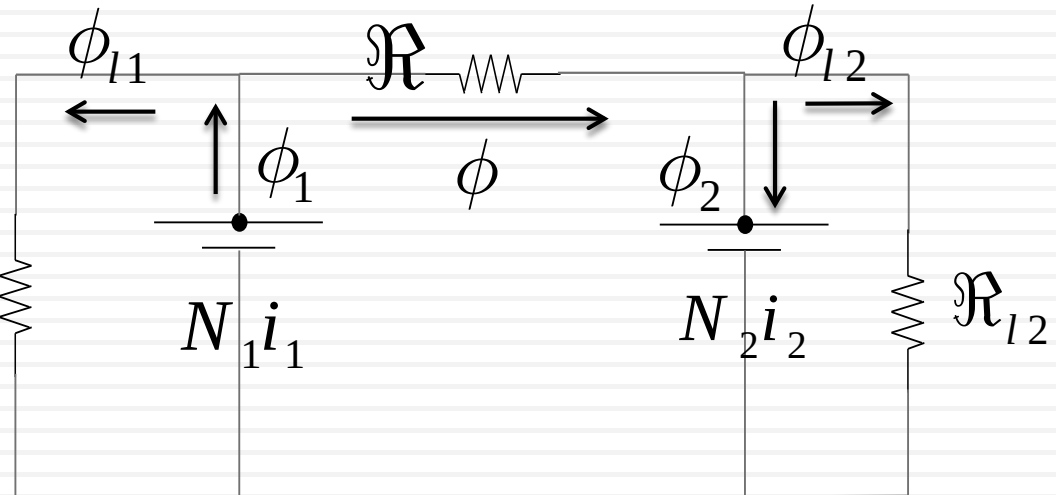
$$\phi_2 = \phi - \phi_{l2} \quad (2)$$

$$N_1 i_1 = \phi_{l1} \mathcal{R}_{l1} \quad (3)$$

$$N_2 i_2 = \phi_{l2} \mathcal{R}_{l2} \quad (4)$$

From (1) & (3)

$$N_1 \frac{d\phi_1}{dt} = N_1 \frac{d\phi}{dt} + N_1 \frac{d\phi_{l1}}{dt} \Rightarrow v_{l1} = N_1 \frac{d\phi_{l1}}{dt} = \frac{N_1^2}{R_{l1}} \frac{di_1}{dt}$$



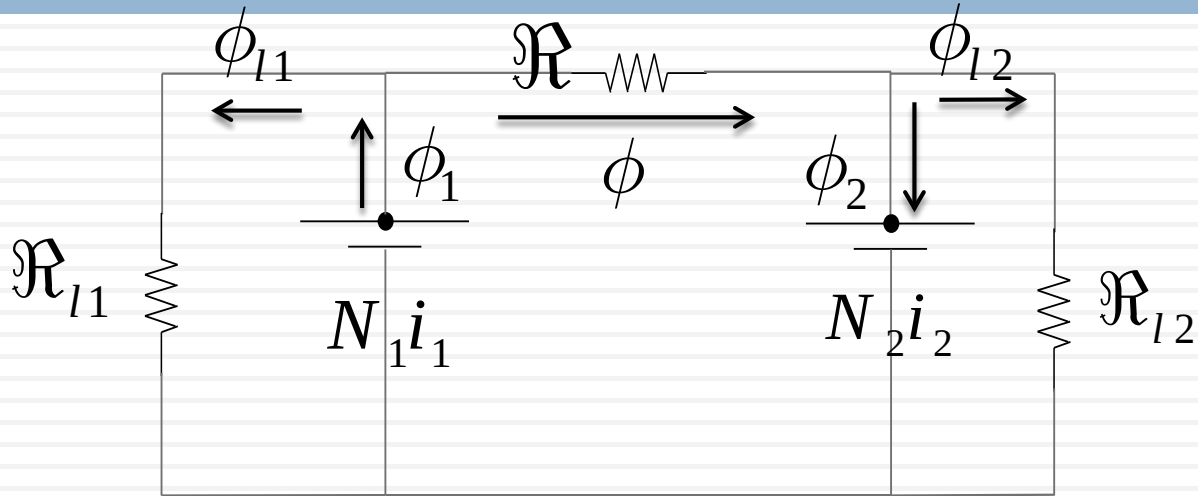
Transformer: Equivalent Circuit

27

Case D:

$$\phi_2 = \phi - \phi_{l2} \quad (2)$$

$$N_2 i_2 = \phi_{l2} \mathcal{R}_{l2} \quad (4)$$



From (2) & (4)

$$N_2 \frac{d\phi_2}{dt} = N_2 \frac{d\phi}{dt} - N_2 \frac{d\phi_{l2}}{dt}$$



$$v_{l2} = N_2 \frac{d\phi_{l2}}{dt} = \frac{N_2^2}{R_{l2}} \frac{di_2}{dt}$$

Transformer: Equivalent Circuit

28

Case D:

let

$$N_1 \frac{d\phi_1}{dt} = N_1 \frac{d\phi}{dt} + N_1 \frac{d\phi_{l1}}{dt}$$

$$N_2 \frac{d\phi_2}{dt} = N_2 \frac{d\phi}{dt} - N_2 \frac{d\phi_{l2}}{dt}$$

$$v_1 = N_1 \frac{d\phi_1}{dt}$$

$$e_1 = N_1 \frac{d\phi}{dt}$$

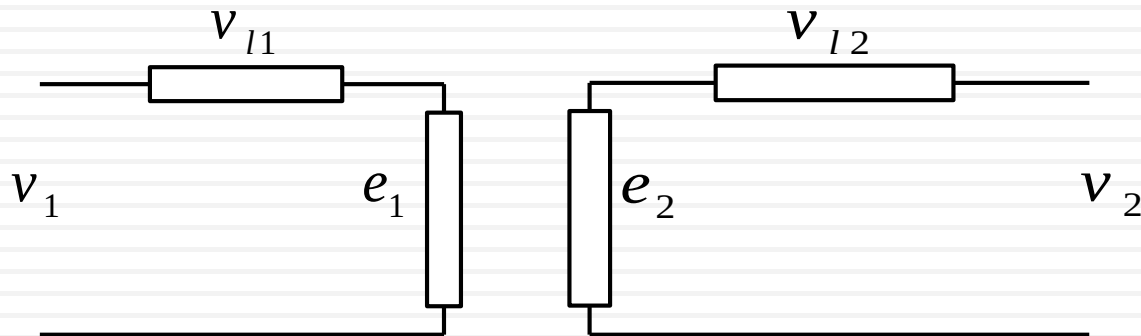
$$v_2 = N_2 \frac{d\phi_2}{dt}$$

$$e_2 = N_2 \frac{d\phi}{dt}$$



$$v_1 = e_1 + v_{l1}$$

$$v_2 = e_2 - v_{l2}$$



Transformer: Equivalent Circuit

29

Case D:

$$v_{l1} = \frac{N_1^2}{R_{l1}} \frac{di_1}{dt}$$

$$v_{l1} = L_{l1} \frac{di_1}{dt}$$

$$v_{l2} = \frac{N_2^2}{R_{l2}} \frac{di_2}{dt}$$

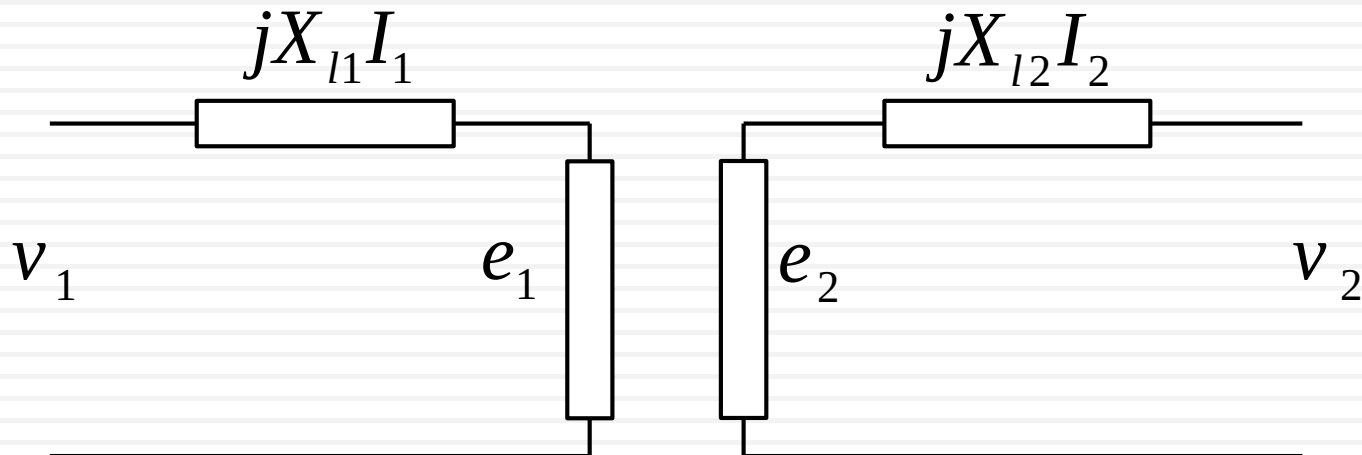
$$v_{l2} = L_{l2} \frac{di_2}{dt}$$

$$V_{l1} = j\omega L_{l1} I_1$$

$$V_{l1} = jX_{l1} I_1$$

$$V_{l2} = j\omega L_{l2} I_2$$

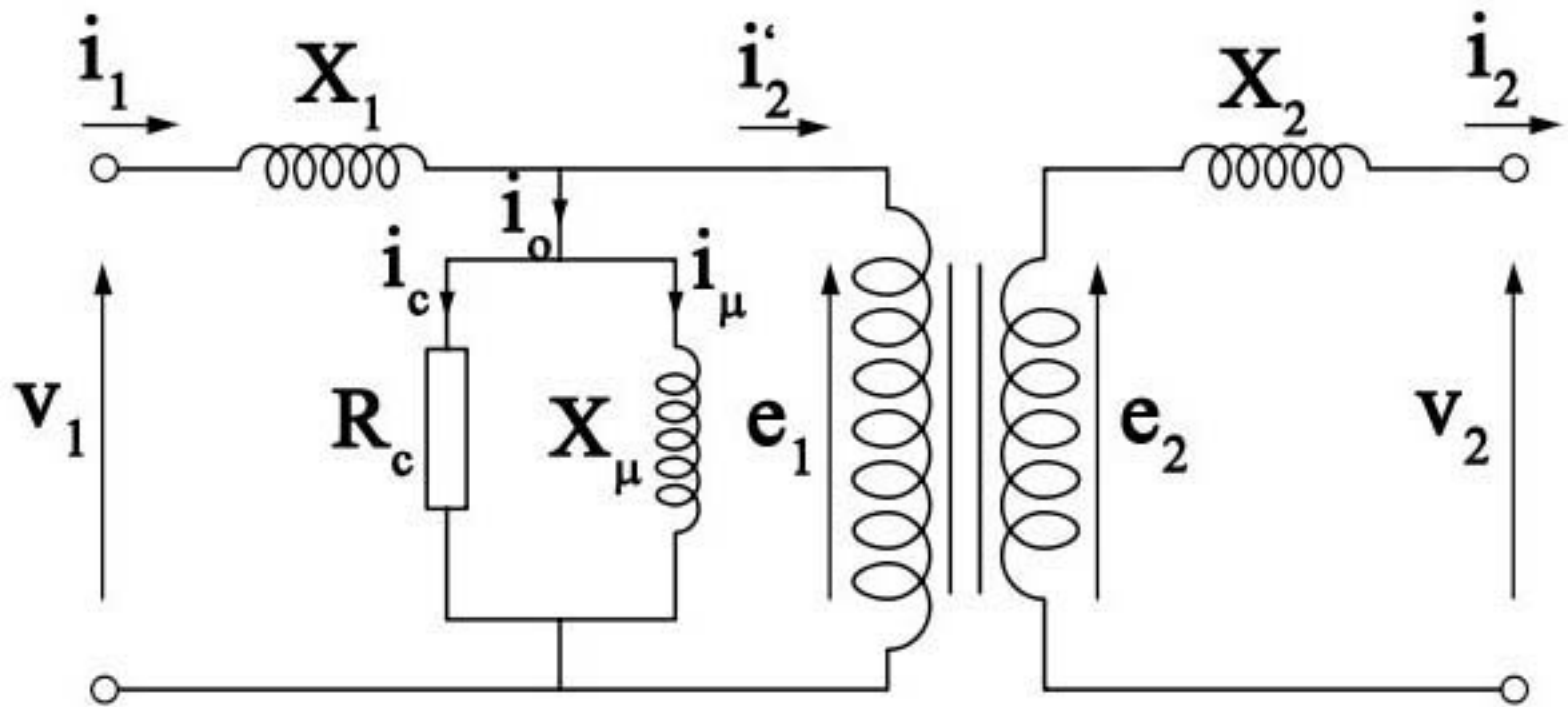
$$V_{l2} = jX_{l2} I_2$$



Transformer: Equivalent Circuit

30

Case D:



Transformer: Equivalent Circuit

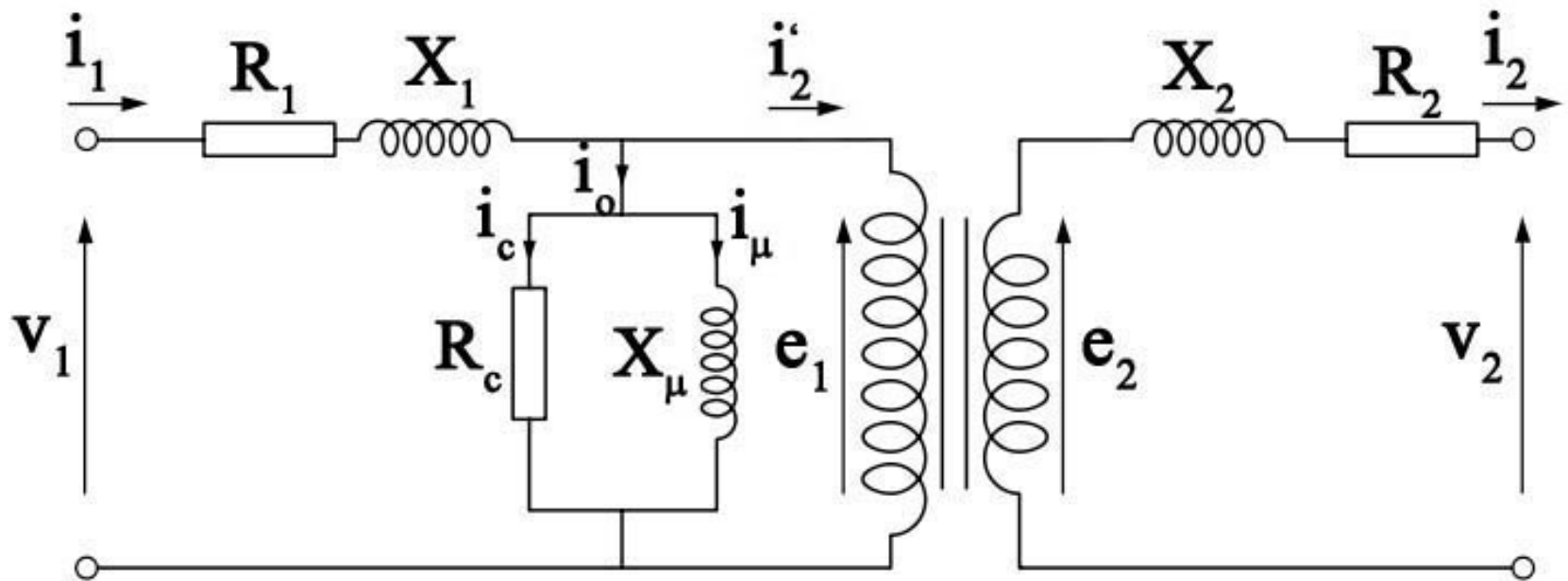
31

Case E: - $R_{\text{coil}} \neq \text{zero}$

- $\mu \neq \infty$

- Leakage flux exists.

- core losses exist.



Exact Equivalent Circuit of the Transformer